# 1.8V Low-Power Open-Drain Output Comparator

#### Features:

- Propagation Delay at 1.8 V<sub>DD</sub>: 56 ns (typical) High-to-Low
- Low Quiescent Current: 100 μA (typical)
- Input Offset Voltage: ±3 mV (typical)
- Rail-to-Rail Input: V<sub>SS</sub> 0.3V to V<sub>DD</sub> + 0.3V
- · Open-Drain Output
- Wide Supply Voltage Range: 1.8V to 5.5V
- · Available in Single, Dual and Quad
- Packages: SC70, SOT-23-5, SOIC, MSOP, TSSOP

### **Typical Applications:**

- · Laptop Computers
- Mobile Phones
- · Hand-held Electronics
- RC Timers
- · Alarm and Monitoring Circuits
- · Window Comparators
- Multivibrators

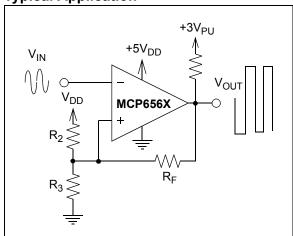
## **Design Aids:**

- · Microchip Advanced Part Selector (MAPS)
- · Analog Demonstration and Evaluation Boards
- · Application Notes
- · SPICE Macro Model

#### **Related Device:**

Push-Pull Output: MCP6561/1R/1U/2/4

#### **Typical Application**



#### **Description:**

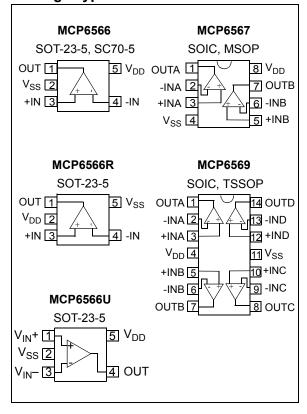
The Microchip MCP6566/6R/6U/7/9 family of opendrain output comparators are offered in single, dual and quad configurations.

These comparators are optimized for low-power 1.8V, single-supply applications with greater than rail-to-rail input operation. The internal input hysteresis eliminates output switching due to internal input noise voltage, reducing current draw.

The open-drain output of the MCP6566/6R/6U/7/9 family requires a pull-up resistor. It supports pull-up voltages that are above and below  $V_{DD}$ , which can be used to level shift. The output toggle frequency can reach a typical of 4 MHz (typical) while limiting supply current surges and dynamic power consumption during switching.

This family operates with single supply voltage of 1.8V to 5.5V while drawing less than 100 µA/comparator of quiescent current (typical).

## **Package Types**



NOTES:

# 1.0 ELECTRICAL CHARACTERISTICS

### 1.1 Maximum Ratings †

V <sub>DD</sub> - V <sub>SS</sub>	6.5V
Open-Drain Output	V <sub>SS</sub> + 10.5V
All other inputs and outputs $\mbox{\ensuremath{V_{SS}}}$	$-0.3V$ to $V_{DD} + 0.3V$
Analog Input (V <sub>IN</sub> ) ††V <sub>SS</sub>	– 1.0V to V <sub>DD</sub> + 1.0V
Difference Input voltage	V <sub>DD</sub> - V <sub>SS</sub>
Output Short Circuit Current	±25 mA
Current at Input Pins	±2 mA
Current at Output and Supply Pins	±50 mA
Storage temperature	65°C to +150°C
Ambient temp. with power applied	40°C to +125°C
Junction temp	+150°C
ESD protection on all pins (HBM/MM)	≥ 4kV/300V

† Notice: Stresses above those listed under "Maximum Ratings" may cause permanent damage to the device. This is a stress rating only and functional operation of the device at those or any other conditions above those indicated in the operational listings of this specification is not implied. Exposure to maximum rating conditions for extended periods may affect device reliability.

†† See Section 4.1.2 "Input Voltage and Current Limits"

#### DC CHARACTERISTICS

**Electrical Characteristics:** Unless otherwise indicated:  $V_{DD}$  = +1.8V to +5.5V,  $V_{SS}$  = GND,  $T_A$  = +25°C,  $V_{IN}$ + =  $V_{DD}$ /2,  $V_{IN}$ - =  $V_{SS}$ , and  $R_{Pull-ID}$  = 20 kΩ to  $V_{PLI}$  =  $V_{DD}$  (see Figure 1-1).

and R <sub>Pull-Up</sub> = 20 ks2 to V <sub>PU</sub> = V <sub>DD</sub> (see	and $R_{Pull-Up} = 20 \text{ kg}$ to $V_{PU} = V_{DD}$ (see Figure 1-1).								
Parameters	Symbol	Min.	Тур.	Max.	Units	Conditions			
Power Supply									
Supply Voltage	$V_{DD}$	1.8	_	5.5	V				
Quiescent Current per comparator	IQ	60	100	130	μΑ	I <sub>OUT</sub> = 0			
Power Supply Rejection Ratio	PSRR	63	70	_	dB	V <sub>CM</sub> = V <sub>SS</sub>			
Input									
Input Offset Voltage	V <sub>OS</sub>	-10	±3	+10	mV	V <sub>CM</sub> = V <sub>SS</sub> (Note 1)			
Input Offset Drift	ΔV <sub>OS</sub> /ΔT	_	±2	_	μV/°C	V <sub>CM</sub> = V <sub>SS</sub>			
Input Offset Current	los	_	±1	_	pA	V <sub>CM</sub> = V <sub>SS</sub>			
Input Bias Current	I <sub>B</sub>	_	1	_	pA	$T_A = +25^{\circ}C, V_{IN^-} = V_{DD}/2$			
		_	60	_	pA	$T_A = +85^{\circ}C, V_{IN^-} = V_{DD}/2$			
		_	1500	5000	pA	$T_A = +125$ °C, $V_{IN^-} = V_{DD}/2$			
Input Hysteresis Voltage	V <sub>HYST</sub>	1.0	_	5.0	mV	V <sub>CM</sub> = V <sub>SS</sub> (Notes 1, 2)			
Input Hysteresis Linear Temp. Co.	TC <sub>1</sub>	_	10	_	μV/°C				
Input Hysteresis Quadratic Temp. Co.	TC <sub>2</sub>	_	0.3	_	μV/°C <sup>2</sup>				
Common-Mode Input Voltage Range	V <sub>CMR</sub>	V <sub>SS</sub> -0.2	_	V <sub>DD</sub> + 0.2	V	V <sub>DD</sub> = 1.8V			
		V <sub>SS</sub> -0.3	_	V <sub>DD</sub> + 0.3	V	V <sub>DD</sub> = 5.5V			
Common-Mode Rejection Ratio	CMRR	54	66	_	dB	$V_{CM} = -0.3V$ to $V_{DD} + 0.3V$ , $V_{DD} = 5.5V$			
		50	63	_	dB	$V_{CM} = V_{DD}/2$ to $V_{DD}+0.3V$ , $V_{DD} = 5.5V$			
		54	65	_	dB	$V_{CM} = -0.3V \text{ to } V_{DD}/2, V_{DD} = 5.5V$			
Common Mode Input Impedance	Z <sub>CM</sub>	_	10 <sup>13</sup>   4	_	$\Omega    pF$				
Differential Input Impedance	Z <sub>DIFF</sub>	_	10 <sup>13</sup>   2	_	Ω  pF				

**Note** 1: The input offset voltage is the center of the input-referred trip points. The input hysteresis is the difference between the input-referred trip points.

- 2:  $V_{HYST}$  at different temperatures is estimated using  $V_{HYST}$  (T<sub>A</sub>) =  $V_{HYST}$  @ +25°C + (T<sub>A</sub> 25°C) TC<sub>1</sub> + (T<sub>A</sub> 25°C)<sup>2</sup> TC<sub>2</sub>.
- 3: Limit the output current to Absolute Maximum Rating of 50 mA.
- 4: The pull-up voltage for the open drain output V<sub>PULL\_UP</sub> can be as high as the absolute maximum rating of 10.5V. In this case, I<sub>OH leak</sub> can be higher than 1 μA (see Figure 2-30).

# DC CHARACTERISTICS (CONTINUED)

**Electrical Characteristics:** Unless otherwise indicated:  $V_{DD}$  = +1.8V to +5.5V,  $V_{SS}$  = GND,  $T_A$  = +25°C,  $V_{IN}$ + =  $V_{DD}$ /2,  $V_{IN}$ - =  $V_{SS}$ , and  $R_{Pull-Up}$  = 20 kΩ to  $V_{PU}$  =  $V_{DD}$  (see Figure 1-1).

Parameters	Symbol	Min.	Тур.	Max.	Units	Conditions		
Push-Pull Output								
Pull-up Voltage	V <sub>PULL_UP</sub>	1.6	_	5.5	V			
High-Level Output Voltage	V <sub>OH</sub>	_	_	V <sub>PULL_UP</sub>	V	(see Figure 1-1) (Notes 3, 4)		
High-Level Output Current Leakage	I <sub>OH_leak</sub>	_	_	1	μA	Note 4		
Low-Level Output Voltage	V <sub>OL</sub>	_	_	0.6	V	$I_{OUT} = 3 \text{ mA/8 mA} @ V_{DD} = 1.8 \text{V/5.5V}$		
Short Circuit Current (Notes 3)	I <sub>SC</sub>	_	±30	_	mA	Not to exceed Absolute Max. Rating		
Output Pin Capacitance	C <sub>OUT</sub>	_	8	_	pF			

- **Note** 1: The input offset voltage is the center of the input-referred trip points. The input hysteresis is the difference between the input-referred trip points.
  - 2:  $V_{HYST}$  at different temperatures is estimated using  $V_{HYST}$  (T<sub>A</sub>) =  $V_{HYST}$  @ +25°C + (T<sub>A</sub> 25°C) TC<sub>1</sub> + (T<sub>A</sub> 25°C)<sup>2</sup> TC<sub>2</sub>.
  - 3: Limit the output current to Absolute Maximum Rating of 50 mA.
  - 4: The pull-up voltage for the open drain output V<sub>PULL UP</sub> can be as high as the absolute maximum rating of 10.5V. In this case, I<sub>OH leak</sub> can be higher than 1 μA (see Figure 2-30).

#### **AC CHARACTERISTICS**

Electrical Characteristics: Unless otherwise indicated,: Unless otherwise indicated,:  $V_{DD}$  = +1.8V to +5.5V,  $V_{SS}$  = GND,  $T_A = +25^{\circ}C$ ,  $V_{IN+} = V_{DD}/2$ ,  $V_{IN-} = V_{SS}$ ,  $R_{Pull-Up} = 20 \text{ k}\Omega$  to  $V_{PU} = V_{DD}$ , and  $C_L = 25 \text{ pf}$  (see Figure 1-1). **Parameters** Symbol Min. Units Conditions Typ. Max. **Propagation Delay** High-to-Low, 100 mV Overdrive 56 80  $V_{CM} = V_{DD}/2, V_{DD} = 1.8V$  $t_{PHL}$ 34 80  $V_{CM} = V_{DD}/2, V_{DD} = 5.5V$ Output Fall Time 20  $t_{F}$ Maximum Toggle Frequency 4 MHz  $V_{DD} = 5.5V$  $f_{\mathsf{TG}}$ 2 MHz  $V_{DD} = 1.8V$  $\mathsf{E}_\mathsf{NI}$ 350 10 Hz to 10 MHz (Note 1) Input Voltage Noise  $\mu V_{P-P}$ 

- Note 1: ENI is based on SPICE simulation.
  - 2: Rise time t<sub>R</sub> and t<sub>PLH</sub> depend on the load (R<sub>L</sub> and C<sub>L</sub>). These specification are valid for the specified load only.

#### TEMPERATURE SPECIFICATIONS

Electrical Characteristics: Unless otherwise indicated: $V_{DD}$ = +1.8v to +5.5v and $V_{SS}$ = GND.									
Parameters	Symbol	Min.	Тур.	Max.	Units	Conditions			
Temperature Ranges									
Specified Temperature Range	T <sub>A</sub>	-40	_	+125	°C				
Operating Temperature Range	T <sub>A</sub>	-40	_	+125	°C				
Storage Temperature Range	T <sub>A</sub>	-65	_	+150	°C				
Thermal Package Resistances	·								
Thermal Resistance, SC70-5	$\theta_{JA}$	_	331	_	°C/W				
Thermal Resistance, SOT-23-5	$\theta_{JA}$	_	201	_	°C/W				
Thermal Resistance, 8L-MSOP	$\theta_{JA}$	_	211	_	°C/W				
Thermal Resistance, 8L-SOIC	$\theta_{JA}$	_	149	_	°C/W				
Thermal Resistance, 14L-SOIC	$\theta_{JA}$	_	91	_	°C/W				
Thermal Resistance, 14L-TSSOP	$\theta_{JA}$	_	100	_	°C/W				

# 1.2 Test Circuit Configuration

This test circuit configuration is used to determine the AC and DC specifications.

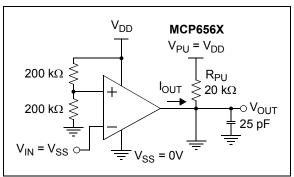


FIGURE 1-1: AC and DC Test Circuit for the Open-Drain Output Comparators.

NOTES:

#### 2.0 TYPICAL PERFORMANCE CURVES

**Note:** The graphs and tables provided following this note are a statistical summary based on a limited number of samples and are provided for informational purposes only. The performance characteristics listed herein are not tested or guaranteed. In some graphs or tables, the data presented may be outside the specified operating range (i.e., outside specified power supply range) and therefore outside the warranted range.

**Note:** Unless otherwise indicated,  $V_{DD}$  = +1.8V to +5.5V,  $V_{SS}$  = GND,  $T_A$  = +25°C,  $V_{IN}$ + =  $V_{DD}$ /2,  $V_{IN}$ - = GND,  $R_L$  = 20 k $\Omega$  to  $V_{PU}$  =  $V_{DD}$ , and  $C_L$  = 25 pF.

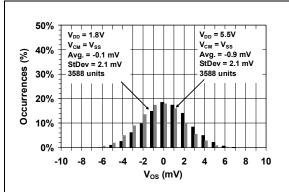


FIGURE 2-1: Input Offset Voltage.

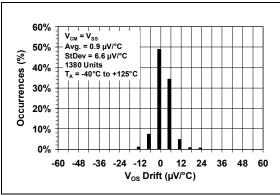


FIGURE 2-2: Input Offset Voltage Drift.

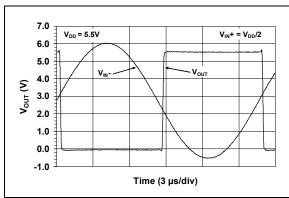


FIGURE 2-3: Input vs. Output Signal, No Phase Reversal.

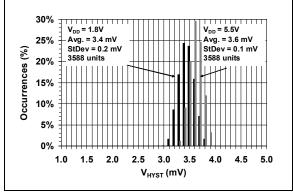


FIGURE 2-4: Input Hysteresis Voltage.

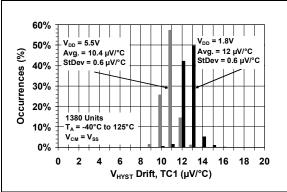


FIGURE 2-5: Input Hysteresis Voltage Drift – Linear Temp. Co. (TC1).

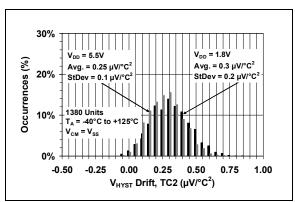


FIGURE 2-6: Input Hysteresis Voltage Drift – Quadratic Temp. Co. (TC2).

**Note:** Unless otherwise indicated,  $V_{DD}$  = +1.8V to +5.5V,  $V_{SS}$  = GND,  $T_A$  = +25°C,  $V_{IN}$ + =  $V_{DD}$ /2,  $V_{IN^-}$  = GND,  $R_L$  = 20 k $\Omega$  to  $V_{PU}$  =  $V_{DD}$ , and  $C_L$  = 25 pF.

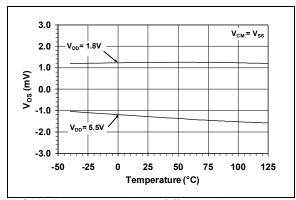


FIGURE 2-7: Temperature.

Input Offset Voltage vs.

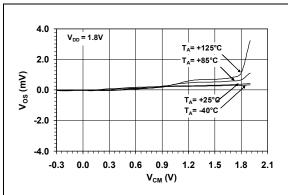


FIGURE 2-8: Input Offset Voltage vs. Common-mode Input Voltage.

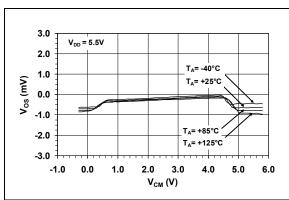
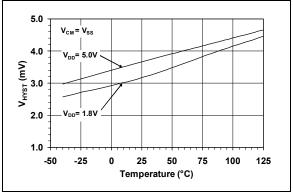


FIGURE 2-9: Input Offset Voltage vs. Common-mode Input Voltage.



**FIGURE 2-10:** Input Hysteresis Voltage vs. Temperature.

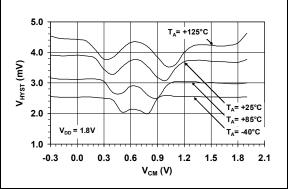


FIGURE 2-11: Input Hysteresis Voltage vs. Common-mode Input Voltage.

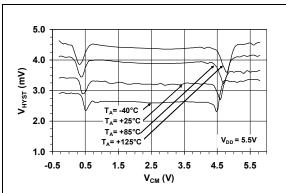


FIGURE 2-12: Input Hysteresis Voltage vs. Common-mode Input Voltage.

**Note:** Unless otherwise indicated,  $V_{DD}$  = +1.8V to +5.5V,  $V_{SS}$  = GND,  $T_A$  = +25°C,  $V_{IN}$ + =  $V_{DD}$ /2,  $V_{IN^-}$  = GND,  $R_L$  = 20 k $\Omega$  to  $V_{PU}$  =  $V_{DD}$ , and  $C_L$  = 25 pF.

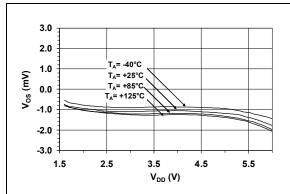


FIGURE 2-13: Input Offset Voltage vs. Supply Voltage vs. Temperature.

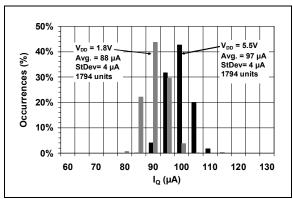
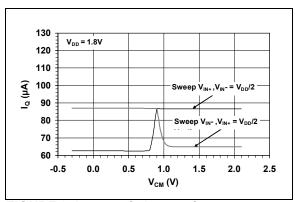


FIGURE 2-14: Quiescent Current.



**FIGURE 2-15:** Quiescent Current vs. Common-mode Input Voltage.

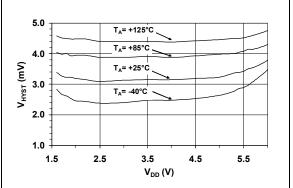


FIGURE 2-16: Input Hysteresis Voltage vs. Supply Voltage vs. Temperature.

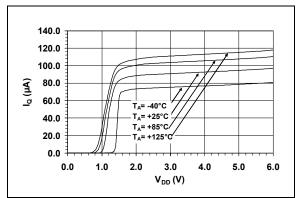


FIGURE 2-17: Quiescent Current vs. Supply Voltage vs. Temperature.

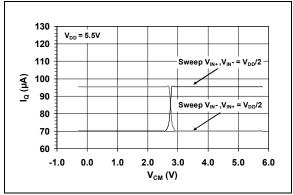
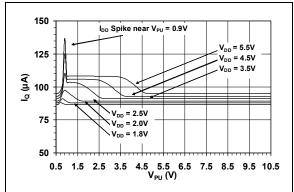


FIGURE 2-18: Quiescent Current vs. Common-mode Input Voltage.

**Note:** Unless otherwise indicated,  $V_{DD}$  = +1.8V to +5.5V,  $V_{SS}$  = GND,  $T_A$  = +25°C,  $V_{IN}$ + =  $V_{DD}$ /2,  $V_{IN^-}$  = GND,  $R_L$  = 20 k $\Omega$  to  $V_{PU}$  =  $V_{DD}$ , and  $C_L$  = 25 pF.



**FIGURE 2-19:** Quies Pull-up Voltage.

Quiescent Current vs.

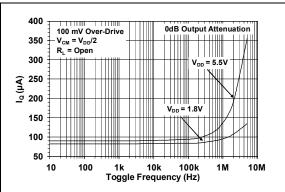


FIGURE 2-20: Quiescent Current vs. Toggle Frequency.

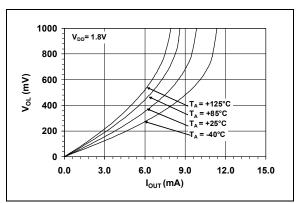


FIGURE 2-21: Output Headroom vs. Output Current.

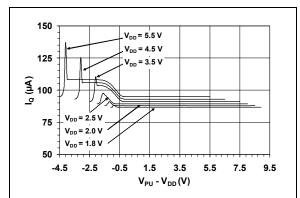


FIGURE 2-22: Quiescent Current vs. Pull-up to Supply Voltage Difference.

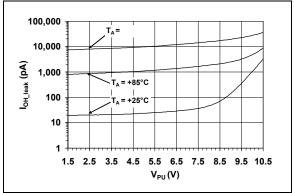


FIGURE 2-23: Output Leakage Current vs. Pull-up Voltage.

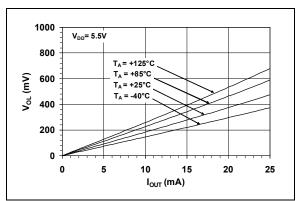


FIGURE 2-24:
Output Current.

Output Headroom vs.

**Note:** Unless otherwise indicated,  $V_{DD}$  = +1.8V to +5.5V,  $V_{SS}$  = GND,  $T_A$  = +25°C,  $V_{IN}$ + =  $V_{DD}$ /2,  $V_{IN^-}$  = GND,  $R_L$  = 20 k $\Omega$  to  $V_{PU}$  =  $V_{DD}$ , and  $C_L$  = 25 pF.

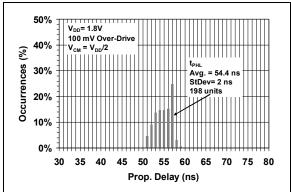


FIGURE 2-25: High-to-Low Propagation Delays.

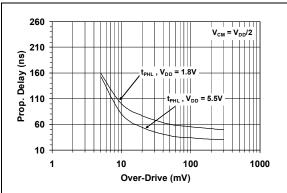
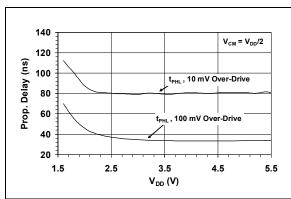


FIGURE 2-26: Propagation Delay vs. Input Over-Drive.



**FIGURE 2-27:** Propagation Delay vs. Supply Voltage.

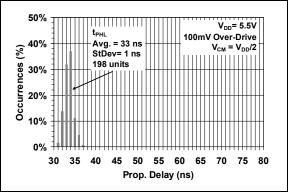
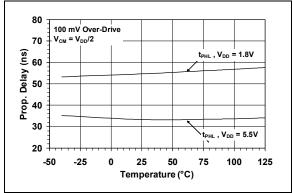


FIGURE 2-28: High-to-Low Propagation Delays.



**FIGURE 2-29:** Propagation Delay vs. Temperature.

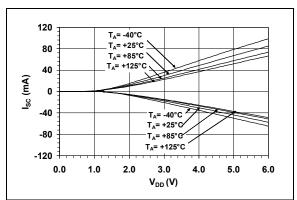
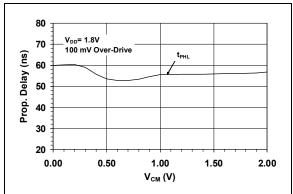


FIGURE 2-30: Short Circuit Current vs. Supply Voltage vs. Temperature.

**Note:** Unless otherwise indicated,  $V_{DD}$  = +1.8V to +5.5V,  $V_{SS}$  = GND,  $T_A$  = +25°C,  $V_{IN}$ + =  $V_{DD}$ /2,  $V_{IN^-}$  = GND,  $R_L$  = 20 k $\Omega$  to  $V_{PU}$  =  $V_{DD}$ , and  $C_L$  = 25 pF.



**FIGURE 2-31:** Propagation Delay vs. Common-mode Input Voltage.

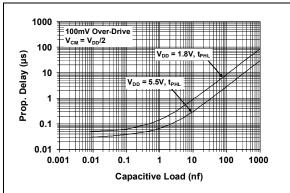
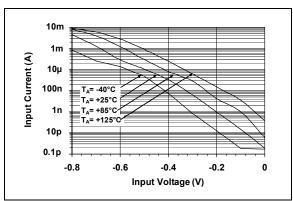
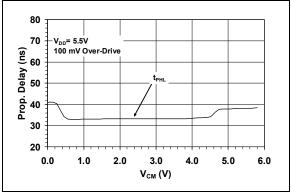


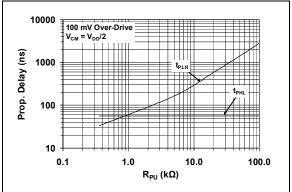
FIGURE 2-32: Propagation Delay vs. Capacitive Load.



**FIGURE 2-33:** Input Bias Current vs. Input Voltage vs. Temperature.



**FIGURE 2-34:** Propagation Delay vs. Common-mode Input Voltage.



**FIGURE 2-35:** Propagation Delay vs. Pull-up Resistor.

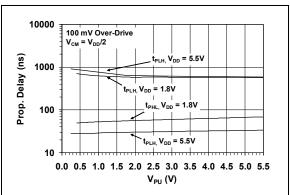
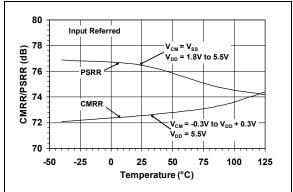


FIGURE 2-36: Propagation Delay vs. Pull-up Voltage.

**Note:** Unless otherwise indicated,  $V_{DD}$  = +1.8V to +5.5V,  $V_{SS}$  = GND,  $T_A$  = +25°C,  $V_{IN}$ + =  $V_{DD}$ /2,  $V_{IN}$ - = GND,  $R_L$  = 20 k $\Omega$  to  $V_{PU}$  =  $V_{DD}$ , and  $C_L$  = 25 pF.



**FIGURE 2-37:** Common-mode Rejection Ratio and Power Supply Rejection Ratio vs. Temperature.

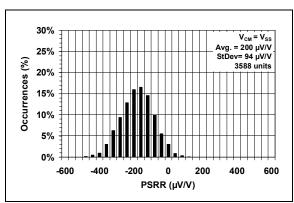


FIGURE 2-38: Power Supply Rejection Ratio (PSRR).

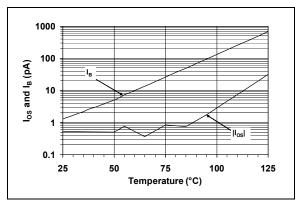


FIGURE 2-39: Input Offset Current and Input Bias Current vs. Temperature.

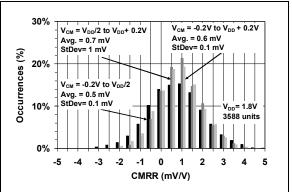
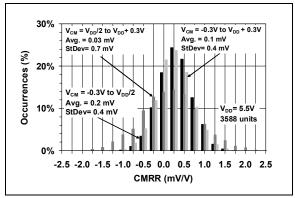


FIGURE 2-40: Common-mode Rejection Ratio (CMRR).



**FIGURE 2-41:** Common-mode Rejection Ratio (CMRR).

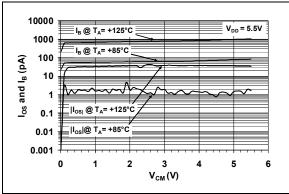


FIGURE 2-42: Input Offset Current and Input Bias Current vs. Common-mode Input Voltage vs. Temperature.

**Note:** Unless otherwise indicated,  $V_{DD}$  = +1.8V to +5.5V,  $V_{SS}$  = GND,  $T_A$  = +25°C,  $V_{IN}$ + =  $V_{DD}$ /2,  $V_{IN}$ - = GND,  $R_L = 20 \text{ k}\Omega$  to  $V_{PU} = V_{DD}$ , and  $C_L = 25 \text{ pF}$ .

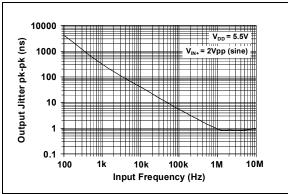


FIGURE 2-43:

Output Jitter vs. Input

#### 3.0 PIN DESCRIPTIONS

Descriptions of the pins are listed in Table 3-1.

TABLE 3-1: PIN FUNCTION TABLE

MCP6566	MCP6566R	MCP6566U	MCP6567	MCP6569		
SC70-5, SOT-23-5	SOT-23-5	SOT-23-5	MSOP, SOIC	SOIC, TSSOP	Symbol	Description
1	1	4	1	1	OUT, OUTA	Digital Output (comparator A)
4	4	3	2	2	V <sub>IN</sub> -, V <sub>INA</sub> -	Inverting Input (comparator A)
3	3	1	3	3	$V_{IN}$ +, $V_{INA}$ +	Non-inverting Input (comparator A)
5	2	5	8	4	$V_{DD}$	Positive Power Supply
_	_	_	5	5	V <sub>INB</sub> +	Non-inverting Input (comparator B)
_	_	_	6	6	V <sub>INB</sub> -	Inverting Input (comparator B)
_	_	_	7	7	OUTB	Digital Output (comparator B)
_	_	_	_	8	OUTC	Digital Output (comparator C)
_	_	_	_	9	V <sub>INC</sub> -	Inverting Input (comparator C)
_	_	_	_	10	V <sub>INC</sub> +	Non-inverting Input (comparator C)
2	5	2	4	11	V <sub>SS</sub>	Negative Power Supply
_	_	_	_	12	V <sub>IND</sub> +	Non-inverting Input (comparator D)
_	_	_	_	13	V <sub>IND</sub> -	Inverting Input (comparator D)
_	_	_	_	14	OUTD	Digital Output (comparator D)

### 3.1 Analog Inputs

The comparator non-inverting and inverting inputs are high-impedance CMOS inputs with low bias currents.

### 3.2 Digital Outputs

The comparator outputs are CMOS, open-drain digital outputs. They are designed to make level shifting and wired-OR easy to implement.

## 3.3 Power Supply ( $V_{SS}$ and $V_{DD}$ )

The positive power supply pin ( $V_{DD}$ ) is 1.8V to 5.5V higher than the negative power supply pin ( $V_{SS}$ ). For normal operation, the other pins are at voltages between  $V_{SS}$  and  $V_{DD}$ .

Typically, these parts are used in a single (positive) supply configuration. In this case,  $V_{SS}$  is connected to ground and  $V_{DD}$  is connected to the supply.  $V_{DD}$  will need a local bypass capacitor (typically 0.01  $\mu F$  to 0.1  $\mu F$ ) within 2 mm of the  $V_{DD}$  pin. These can share a bulk capacitor with nearby analog parts (within 100 mm), but it is not required.

NOTES:

#### 4.0 APPLICATIONS INFORMATION

The MCP6566/6R/6U/7/9 family of open-drain output comparators are fabricated on Microchip's state-of-the-art CMOS process. They are suitable for a wide range of high speed applications that require low-power consumption.

### 4.1 Comparator Inputs

#### 4.1.1 NORMAL OPERATION

The input stage of this family of devices uses two different input stages in parallel. This configuration provides three regions of operation, one operates at low input voltages, one at high input voltages, and one at mid input voltage. With this topology, the input voltage range is 0.3V above  $V_{DD}$  and 0.3V below  $V_{SS}$ , while providing low offset voltage throughout the common mode range. The input offset voltage is measured at both  $V_{SS}$  - 0.3V and  $V_{DD}$  + 0.3V to ensure proper operation.

The MCP6566/6R/6U/7/9 family has an internally-set hysteresis  $V_{HYST}$ , which is small enough to maintain input offset accuracy and large enough to eliminate output chattering caused by the comparator's own input noise voltage  $E_{NI}.$  Figure 4-1 depicts this behavior. Input offset voltage ( $V_{OS}$ ) is the center (average) of the (input-referred) low-high and high-low trip points. Input hysteresis voltage ( $V_{HYST}$ ) is the difference between the same trip points.

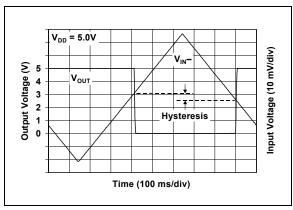


FIGURE 4-1: The MCP6566/6R/6U/7/9 comparators' internal hysteresis eliminates output chatter caused by input noise voltage.

# 4.1.2 INPUT VOLTAGE AND CURRENT LIMITS

The ESD protection on the inputs can be depicted as shown in Figure 4-2. This structure was chosen to protect the input transistors, and to minimize input bias current ( $I_B$ ). The input ESD diodes clamp the inputs when they try to go more than one diode drop below  $V_{SS}$ . They also clamp any voltages that go too far above  $V_{DD}$ . The diodes' breakdown voltage is high enough to allow normal operation, but low enough to bypass ESD events within the specified limits.

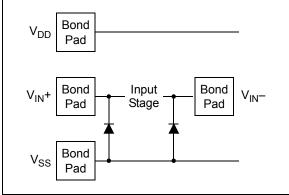


FIGURE 4-2: Simplified Analog Input ESD Structures.

In order to prevent damage and/or improper operation of these amplifiers, the circuits they are in must limit the currents (and voltages) at the  $V_{IN}^{+}$  and  $V_{IN}^{-}$  pins (see Section 1.1 "Maximum Ratings †" at the beginning of Section 1.0 "Electrical Characteristics"). Figure 4-3 shows the recommended approach to protecting these inputs. The internal ESD diodes prevent the input pins  $(V_{IN}^{+}$  and  $V_{IN}^{-})$  from going too far below ground, and the resistors  $R_1$  and  $R_2$  limit the possible current drawn out of the input pin. Diodes  $D_1$  and  $D_2$  prevent the input pin  $(V_{IN}^{+}$  and  $V_{IN}^{-})$  from going too far above  $V_{DD}^{-}$ . When implemented as shown, resistors  $R_1$  and  $R_2$  also limit the current through  $D_1$  and  $D_2$ .

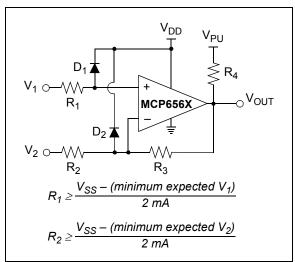


FIGURE 4-3: Protecting the Analog Inputs.

It is also possible to connect the diodes to the left of the resistors  $\mathsf{R}_1$  and  $\mathsf{R}_2.$  In this case, the currents through the diodes  $\mathsf{D}_1$  and  $\mathsf{D}_2$  need to be limited by some other mechanism. The resistor then serves as in-rush current limiter; the DC current into the input pins (V<sub>IN</sub>+ and V<sub>IN</sub>-) should be very small.

A significant amount of current can flow out of the inputs when the common mode voltage ( $V_{CM}$ ) is below ground ( $V_{SS}$ ); see Figure 4-3. Applications that are high-impedance may need to limit the usable voltage range.

#### 4.1.3 PHASE REVERSAL

The MCP6566/6R/6U/7/9 comparator family uses CMOS transistors at the input. They are designed to prevent phase inversion when the input pins exceed the supply voltages. Figure 2-3 shows an input voltage exceeding both supplies with no resulting phase inversion.

#### 4.2 Open-Drain Output

The open-drain output is designed to make level-shifting and wired-OR logic easy to implement. The output stage minimizes switching current (shoot-through current from supply-to-supply) when the output changes state. See Figures 2-15, 2-18, 2-35 and 2-36, for more information.

## 4.3 Externally Set Hysteresis

Greater flexibility in selecting hysteresis (or input trip points) is achieved by using external resistors. Hysteresis reduces output chattering when one input is slowly moving past the other. It also helps in systems in which it is best not to cycle between high and low states too frequently (e.g., air conditioner thermostatic control). Output chatter also increases the dynamic supply current.

#### 4.3.1 NON-INVERTING CIRCUIT

Figure 4-4 shows a non-inverting circuit for single-supply applications using just two resistors. The resulting hysteresis diagram is shown in Figure 4-5.

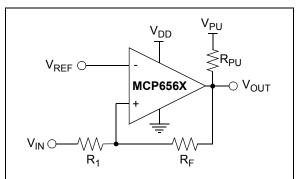


FIGURE 4-4: Non-Inverting Circuit with Hysteresis for Single-Supply.

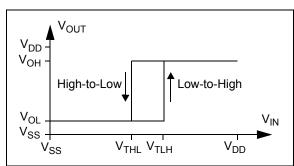


FIGURE 4-5: Hysteresis Diagram for the Non-Inverting Circuit.

The trip points for Figures 4-4 and 4-5 are:

#### **EQUATION 4-1:**

$$V_{TLH} = V_{REF} \left( 1 + \frac{R_I}{R_F} \right) - V_{OL} \left( \frac{R_I}{R_F} \right)$$
$$V_{THL} = V_{REF} \left( 1 + \frac{R_I}{R_F} \right) - V_{OH} \left( \frac{R_I}{R_F} \right)$$

Where:

 $V_{TLH}$  = trip voltage from low-to-high  $V_{THL}$  = trip voltage from high-to-low

#### 4.3.2 INVERTING CIRCUIT

Figure 4-6 shows an inverting circuit for single-supply using three resistors. The resulting hysteresis diagram is shown in Figure 4-7.

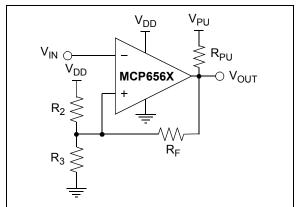


FIGURE 4-6: Inverting Circuit with Hysteresis.

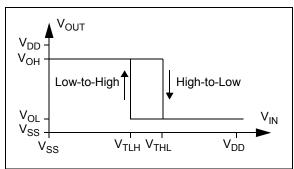


FIGURE 4-7: Hysteresis Diagram for the Inverting Circuit.

In order to determine the trip voltages (V<sub>THL</sub> and V<sub>TLH</sub>) for the circuit shown in Figure 4-6, R<sub>2</sub> and R<sub>3</sub> can be simplified to the Thevenin equivalent circuit with respect to V<sub>DD</sub>, as shown in Figure 4-8.

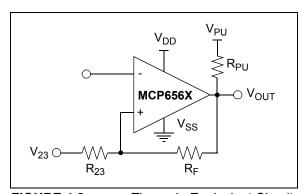


FIGURE 4-8: Thevenin Equivalent Circuit.

Where:

$$R_{23} = \frac{R_2 R_3}{R_2 + R_3}$$

$$V_{23} = \frac{R_3}{R_2 + R_3} \times V_{DD}$$

Using this simplified circuit, the trip voltage can be calculated using the following equation:

#### **EQUATION 4-2:**

$$V_{THL} = V_{OH} \left( \frac{R_{23}}{R_{23} + R_F} \right) + V_{23} \left( \frac{R_F}{R_{23} + R_F} \right)$$

$$V_{TLH} = V_{OL} \left( \frac{R_{23}}{R_{23} + R_F} \right) + V_{23} \left( \frac{R_F}{R_{23} + R_F} \right)$$

Where:

 $V_{TLH}$  = trip voltage from low-to-high  $V_{THL}$  = trip voltage from high-to-low

Figure 2-21 and Figure 2-24 can be used to determine typical values for  $V_{OH}$  and  $V_{OL}$ .

### 4.4 Bypass Capacitors

With this family of comparators, the power supply pin ( $V_{DD}$  for single supply) should have a local bypass capacitor (i.e., 0.01  $\mu$ F to 0.1  $\mu$ F) within 2 mm for good edge rate performance.

#### 4.5 Capacitive Loads

Reasonable capacitive loads (e.g., logic gates) have little impact on propagation delay (see Figure 2-32). The supply current increases with increasing toggle frequency (Figure 2-20), especially with higher capacitive loads. The output slew rate and propagation delay performance will be reduced with higher capacitive loads.

#### 4.6 PCB Surface Leakage

In applications where low input bias current is critical, PCB (Printed Circuit Board) surface leakage effects need to be considered. Surface leakage is caused by humidity, dust or other contamination on the board. Under low humidity conditions, a typical resistance between nearby traces is  $10^{12}\Omega$ . A 5V difference would cause 5 pA of current to flow. This is greater than the MCP6566/6R/6U/7/9 family's bias current at +25°C (1 pA, typical).

The easiest way to reduce surface leakage is to use a guard ring around sensitive pins (or traces). The guard ring is biased at the same voltage as the sensitive pin. An example of this type of layout is shown in Figure 4-9.

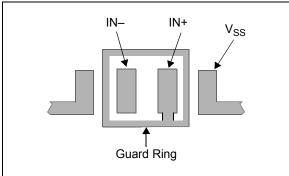


FIGURE 4-9: Example Guard Ring Layout for Inverting Circuit.

- 1. Inverting Configuration (Figures 4-6 and 4-9):
  - a) Connect the guard ring to the non-inverting input pin ( $V_{IN}$ +). This biases the guard ring to the same reference voltage as the comparator (e.g.,  $V_{DD}/2$  or ground).
  - b) Connect the inverting pin ( $V_{IN}$ –) to the input pad without touching the guard ring.
- 2. Non-inverting Configuration (Figure 4-4):
  - a) Connect the non-inverting pin (V<sub>IN</sub>+) to the input pad without touching the guard ring.
  - b) Connect the guard ring to the inverting input pin (V<sub>IN</sub>-).

#### 4.7 PCB Layout Technique

When designing the PCB layout, it is critical to note that analog and digital signal traces are adequately separated to prevent signal coupling. If the comparator output trace is at close proximity to the input traces, then large output voltage changes from, V<sub>SS</sub> to V<sub>DD</sub> or visa versa, may couple to the inputs and cause the device output to oscillate. To prevent such oscillation, the output traces must be routed away from the input pins. The SC70-5 and SOT-23-5 are relatively immune because the output pin OUT (pin 1) is separated by the power pin  $V_{DD}/V_{SS}$  (pin 2) from the input pin +IN (as long as the analog and digital traces remain separated throughout the PCB). However, the pinouts for the dual and quad packages (SOIC, MSOP, TSSOP) have OUT and -IN pins (pin 1 and 2) close to each other. The recommended layout for these packages is shown in Figure 4-10.

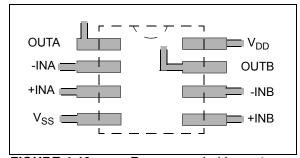


FIGURE 4-10: Recommended Layout.

# 4.8 Unused Comparators

An unused amplifier in a quad package (MCP6569) should be configured as shown in Figure 4-11. This circuit prevents the output from toggling and causing crosstalk. It uses the minimum number of components and draws minimal current (see Figure 2-15 and Figure 2-15).

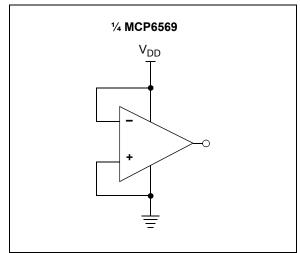


FIGURE 4-11: Unused Comparators.

### 4.9 Typical Applications

#### 4.9.1 PRECISE COMPARATOR

Some applications require higher DC precision. An easy way to solve this problem is to use an amplifier (such as the MCP6291) to gain-up the input signal before it reaches the comparator. Figure 4-12 shows an example of this approach.

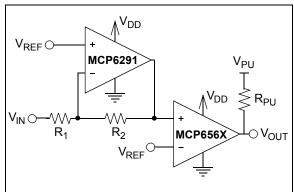


FIGURE 4-12: Precise Inverting Comparator.

#### 4.9.2 WINDOWED COMPARATOR

Figure 4-13 shows one approach to designing a windowed comparator. The AND gate produces a logic '1' when the input voltage is between  $V_{RB}$  and  $V_{RT}$  (where  $V_{RT} > V_{RB}$ ).

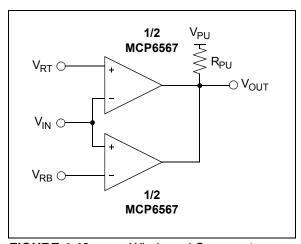


FIGURE 4-13: Windowed Comparator.

#### 4.9.3 BISTABLE MULTIVIBRATOR

A simple bistable multivibrator design is shown in Figure 4-14.  $V_{REF}$  needs to be between the power supplies ( $V_{SS}$  = GND and  $V_{DD})$  to achieve oscillation. The output duty cycle changes with  $V_{REF}$ 

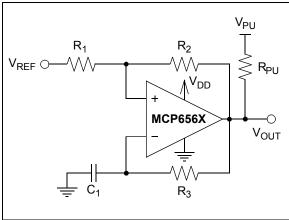


FIGURE 4-14: Bistable Multivibrator.

NOTES:

#### 5.0 DESIGN AIDS

# 5.1 Microchip Advanced Part Selector (MAPS)

MAPS is a software tool that helps semiconductor professionals efficiently identify Microchip devices that fit a particular design requirement.

Available at no cost from the Microchip web site at www.microchip.com/ maps, the MAPS is an overall selection tool for Microchip's product portfolio that includes Analog, Memory, MCU and DSC devices. Using this tool you can define a filter to sort features for a parametric search of devices and export side-by-side technical comparison reports. Helpful links are also provided for data sheets, purchase, and sampling of Microchip parts.

# 5.2 Analog Demonstration and Evaluation Boards

Microchip offers a broad spectrum of Analog Demonstration and Evaluation Boards that are designed to help you achieve faster time to market.

For a listing of these boards and their corresponding user's guides and technical information, visit the Microchip web site at <a href="https://www.microchip.com/analogtools">www.microchip.com/analogtools</a>. Three of the boards that are especially useful are:

- 8-Pin SOIC/MSOP/TSSOP/DIP Evaluation Board, P/N SOIC8EV
- 14-Pin SOIC/TSSOP/DIP Evaluation Board, P/N SOIC14EV
- 5/6-Pin SOT23 Evaluation Board, P/N VSUPEV2

### 5.3 Application Notes

The following Microchip Application Note is available on the Microchip web site at <a href="https://www.microchip.com">www.microchip.com</a> and is recommended as a supplemental reference resource:

AN895 – "Oscillator Circuits for RTD Temperature Sensors" (DS00000895).

#### 5.4 SPICE Macro Model

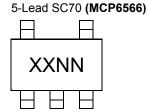
The latest SPICE macro model for the MCP6566/7/9 op amp is available on the Microchip web site at www.microchip.com. The model was written and tested in the official Cadence® (OrCAD®) PSpice®. For the other simulators, translation may be required.

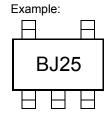
The model covers a wide aspect of the comparator's electrical specifications. Not only does the model cover voltage, current and resistance of the comparator, but it also covers the temperature and the noise effects on the behavior of the comparator. The model has not been verified outside of the specification range listed in the comparator data sheet. The model behaviors under these conditions cannot ensure it will match the actual comparator performance. Moreover, the model is intended to be an initial design tool. Bench testing is a very important part of any design and cannot be replaced with simulations. Also, simulation results using this macro model need to be validated by comparing them to the data sheet specifications and characteristic curves.

NOTES:

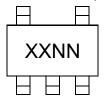
#### 6.0 PACKAGING INFORMATION

#### 6.1 **Package Marking Information**





#### 5-Lead SOT-23 (MCP6566, MCP6566R)



Device	Code			
MCP6566T	JYNN			
MCP6566RT	JZNN			
MCP6566UT WLNN				
Note: Applies to 5-Lead SOT-23.				

Example:



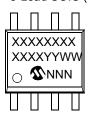
8-Lead MSOP (MCP6567)







8-Lead SOIC (150 mil) (MCP6567)







Legend: XX...X Customer-specific information

Year code (last digit of calendar year) ΥY Year code (last 2 digits of calendar year) WW Week code (week of January 1 is week '01')

NNN Alphanumeric traceability code

(e3) Pb-free JEDEC designator for Matte Tin (Sn)

This package is Pb-free. The Pb-free JEDEC designator (@3)

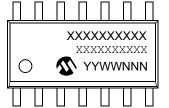
can be found on the outer packaging for this package.

In the event the full Microchip part number cannot be marked on one line, it will Note: be carried over to the next line, thus limiting the number of available

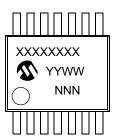
characters for customer-specific information.

## **Package Marking Information (Continued)**

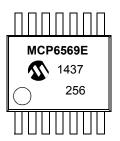




14-Lead TSSOP (MCP6569)

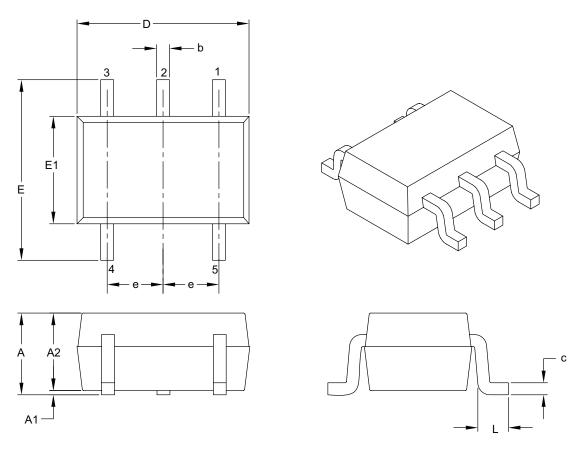






# 5-Lead Plastic Small Outline Transistor (LT) [SC70]

**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



	MILLIMETERS					
Dimensio	n Limits	MIN	NOM	MAX		
Number of Pins	N		5			
Pitch	е		0.65 BSC			
Overall Height	Α	0.80	_	1.10		
Molded Package Thickness	A2	0.80	_	1.00		
Standoff	A1	0.00	_	0.10		
Overall Width	Е	1.80	2.10	2.40		
Molded Package Width	E1	1.15	1.25	1.35		
Overall Length	D	1.80	2.00	2.25		
Foot Length	L	0.10	0.20	0.46		
Lead Thickness	С	0.08	_	0.26		
Lead Width	b	0.15	_	0.40		

#### Notes:

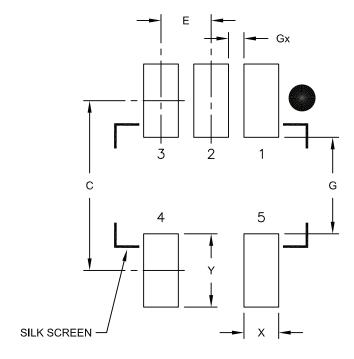
- 1. Dimensions D and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed 0.127 mm per side.
- 2. Dimensioning and tolerancing per ASME Y14.5M.

BSC: Basic Dimension. Theoretically exact value shown without tolerances.

Microchip Technology Drawing C04-061B

# 5-Lead Plastic Small Outline Transistor (LT) [SC70]

**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



RECOMMENDED LAND PATTERN

	Units	MILLIMETERS			
Dimension	Limits	MIN	NOM	MAX	
Contact Pitch	E		0.65 BSC		
Contact Pad Spacing	С		2.20		
Contact Pad Width	Х			0.45	
Contact Pad Length	Υ			0.95	
Distance Between Pads	G	1.25			
Distance Between Pads	Gx	0.20			

#### Notes:

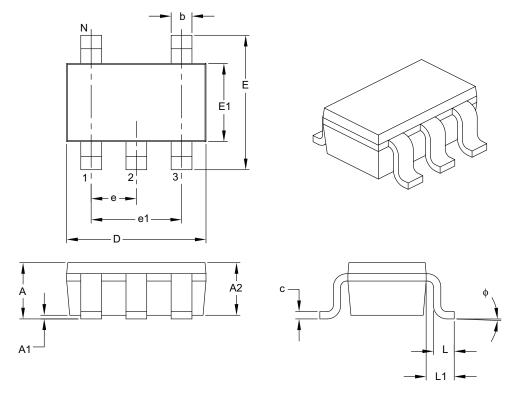
1. Dimensioning and tolerancing per ASME Y14.5M

BSC: Basic Dimension. Theoretically exact value shown without tolerances.

Microchip Technology Drawing No. C04-2061A

# 5-Lead Plastic Small Outline Transistor (OT) [SOT-23]

**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



	Units			3
Dimensio	n Limits	MIN	NOM	MAX
Number of Pins	N		5	
Lead Pitch	е		0.95 BSC	
Outside Lead Pitch	e1		1.90 BSC	
Overall Height	Α	0.90	_	1.45
Molded Package Thickness	A2	0.89	_	1.30
Standoff	A1	0.00	_	0.15
Overall Width	Е	2.20	_	3.20
Molded Package Width	E1	1.30	_	1.80
Overall Length	D	2.70	_	3.10
Foot Length	L	0.10	_	0.60
Footprint	L1	0.35	_	0.80
Foot Angle	ф	0°	_	30°
Lead Thickness	С	0.08	_	0.26
Lead Width	b	0.20	_	0.51

#### Notes:

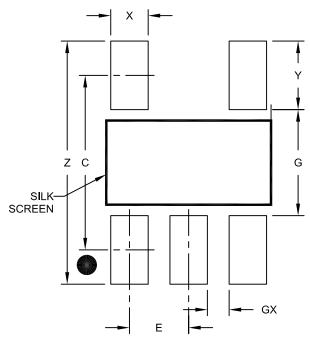
- 1. Dimensions D and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed 0.127 mm per side.
- 2. Dimensioning and tolerancing per ASME Y14.5M.

BSC: Basic Dimension. Theoretically exact value shown without tolerances.

Microchip Technology Drawing C04-091B

# 5-Lead Plastic Small Outline Transistor (OT) [SOT-23]

**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



**RECOMMENDED LAND PATTERN** 

	MILLIMETERS			
Dimension	Dimension Limits		NOM	MAX
Contact Pitch	E			
Contact Pad Spacing	С		2.80	
Contact Pad Width (X5)	Х			0.60
Contact Pad Length (X5)	Υ			1.10
Distance Between Pads	G	1.70		
Distance Between Pads	GX	0.35		
Overall Width	Z			3.90

#### Notes

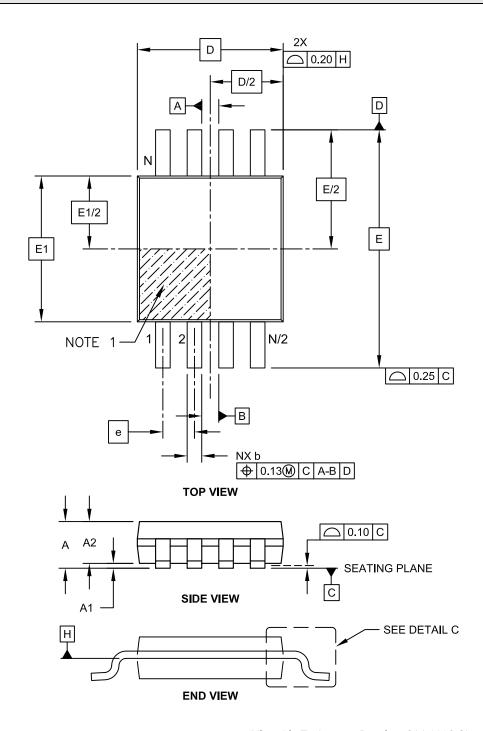
1. Dimensioning and tolerancing per ASME Y14.5M

BSC: Basic Dimension. Theoretically exact value shown without tolerances.

Microchip Technology Drawing No. C04-2091A

# 8-Lead Plastic Micro Small Outline Package (MS) [MSOP]

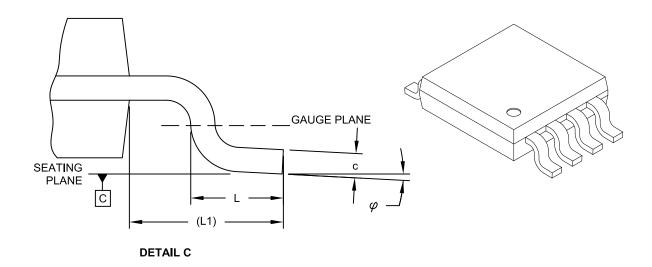
**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



Microchip Technology Drawing C04-111C Sheet 1 of 2

## 8-Lead Plastic Micro Small Outline Package (MS) [MSOP]

**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



	MILLIMETERS				
Dimension Limits		MIN	NOM	MAX	
Number of Pins	N		8		
Pitch	е		0.65 BSC		
Overall Height	Α	-	-	1.10	
Molded Package Thickness	A2	0.75	0.85	0.95	
Standoff	A1	0.00	-	0.15	
Overall Width	E	4.90 BSC			
Molded Package Width	E1	3.00 BSC			
Overall Length	D		3.00 BSC		
Foot Length	L	0.40	0.60	0.80	
Footprint	L1	0.95 REF			
Foot Angle	φ	0°	-	8°	
Lead Thickness	С	0.08	-	0.23	
Lead Width	b	0.22	-	0.40	

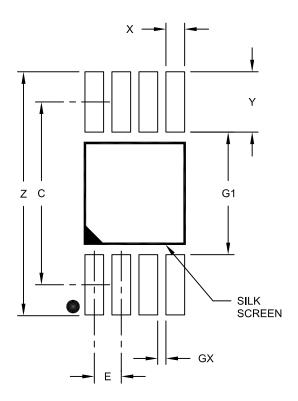
#### Notes:

- 1. Pin 1 visual index feature may vary, but must be located within the hatched area.
- 2. Dimensions D and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed 0.15mm per side.
- 3. Dimensioning and tolerancing per ASME Y14.5M.
  - BSC: Basic Dimension. Theoretically exact value shown without tolerances.
  - REF: Reference Dimension, usually without tolerance, for information purposes only.

Microchip Technology Drawing C04-111C Sheet 2 of 2

## 8-Lead Plastic Micro Small Outline Package (MS) [MSOP]

**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



RECOMMENDED LAND PATTERN

	Units			S
Dimension	Limits	MIN	NOM	MAX
Contact Pitch	Е		0.65 BSC	
Contact Pad Spacing	С		4.40	·
Overall Width	Z			5.85
Contact Pad Width (X8)	X1			0.45
Contact Pad Length (X8)	Y1			1.45
Distance Between Pads	G1	2.95		
Distance Between Pads	GX	0.20		

#### Notes:

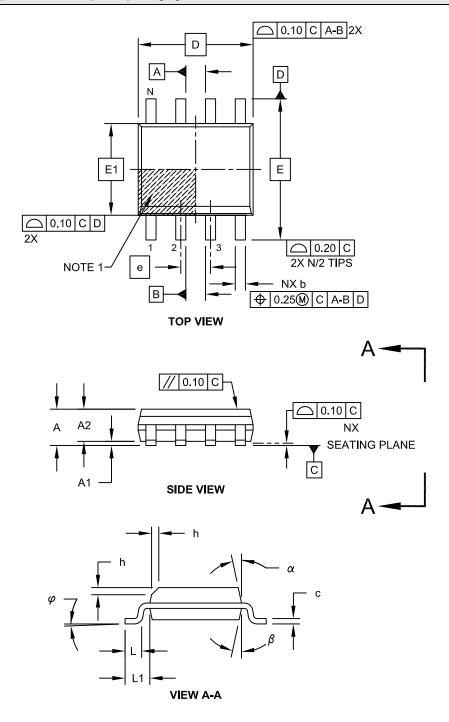
1. Dimensioning and tolerancing per ASME Y14.5M

BSC: Basic Dimension. Theoretically exact value shown without tolerances.

Microchip Technology Drawing No. C04-2111A

## 8-Lead Plastic Small Outline (SN) - Narrow, 3.90 mm Body [SOIC]

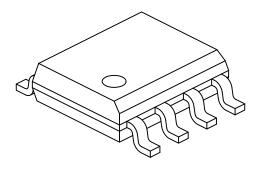
**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



Microchip Technology Drawing No. C04-057C Sheet 1 of 2

# 8-Lead Plastic Small Outline (SN) - Narrow, 3.90 mm Body [SOIC]

**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



Units		MILLIMETERS			
Dimension Limits		MIN	NOM	MAX	
Number of Pins	N	8			
Pitch	е	1.27 BSC			
Overall Height	Α	ı	ı	1.75	
Molded Package Thickness	A2	1.25	ı	-	
Standoff §	A1	0.10	ı	0.25	
Overall Width	Е	6.00 BSC			
Molded Package Width	E1	3.90 BSC			
Overall Length	D	4.90 BSC			
Chamfer (Optional)	h	0.25	ı	0.50	
Foot Length	L	0.40	ı	1.27	
Footprint	L1	1.04 REF			
Foot Angle	$\varphi$	0°	ı	8°	
Lead Thickness	С	0.17	ı	0.25	
Lead Width	b	0.31	ı	0.51	
Mold Draft Angle Top	α	5°	=	15°	
Mold Draft Angle Bottom	β	5°	-	15°	

#### Notes:

- 1. Pin 1 visual index feature may vary, but must be located within the hatched area.
- 2. § Significant Characteristic
- 3. Dimensions D and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed 0.15mm per side.
- 4. Dimensioning and tolerancing per ASME Y14.5M

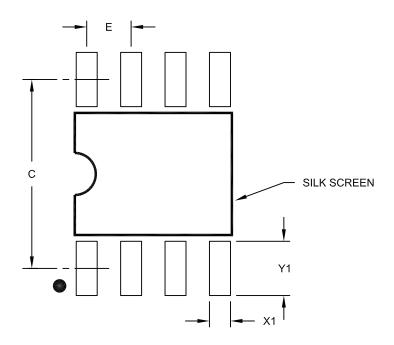
BSC: Basic Dimension. Theoretically exact value shown without tolerances.

REF: Reference Dimension, usually without tolerance, for information purposes only.

Microchip Technology Drawing No. C04-057C Sheet 2 of 2  $\,$ 

# 8-Lead Plastic Small Outline (SN) - Narrow, 3.90 mm Body [SOIC]

**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



RECOMMENDED LAND PATTERN

	Units	MILLIMETERS		
Dimension Limits		MIN	NOM	MAX
Contact Pitch	E	1.27 BSC		
Contact Pad Spacing	С		5.40	
Contact Pad Width (X8)	X1			0.60
Contact Pad Length (X8)	Y1			1.55

#### Notes:

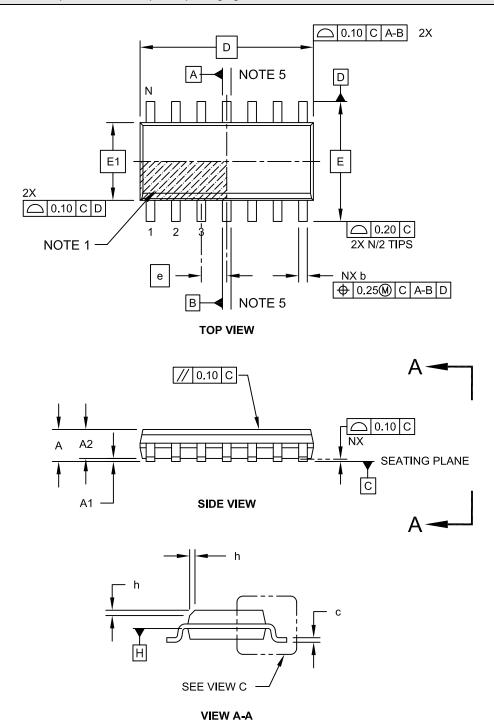
1. Dimensioning and tolerancing per ASME Y14.5M

BSC: Basic Dimension. Theoretically exact value shown without tolerances.

Microchip Technology Drawing No. C04-2057A

#### 14-Lead Plastic Small Outline (SL) - Narrow, 3.90 mm Body [SOIC]

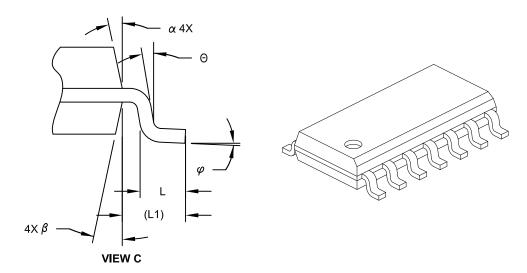
**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



Microchip Technology Drawing No. C04-065C Sheet 1 of 2

#### 14-Lead Plastic Small Outline (SL) - Narrow, 3.90 mm Body [SOIC]

**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



Units		MILLIMETERS		
Dimension Limits		MIN	NOM	MAX
Number of Pins	N	14		
Pitch	е	1.27 BSC		
Overall Height	Α	-	-	1.75
Molded Package Thickness	A2	1.25	-	-
Standoff §	A1	0.10	-	0.25
Overall Width	Е	6.00 BSC		
Molded Package Width	E1	3.90 BSC		
Overall Length	D	8.65 BSC		
Chamfer (Optional)	h	0.25 - 0.50		
Foot Length	L	0.40	-	1.27
Footprint	L1	1.04 REF		
Lead Angle	Θ	0°	-	-
Foot Angle	φ	0°	-	8°
Lead Thickness	С	0.10	-	0.25
Lead Width	b	0.31	-	0.51
Mold Draft Angle Top	α	5°	_	15°
Mold Draft Angle Bottom	β	5°	-	15°

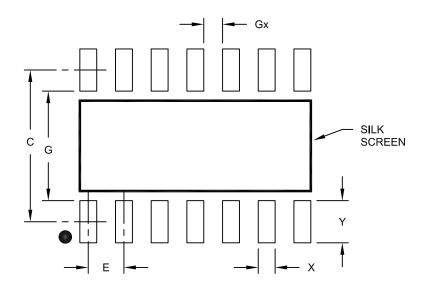
#### Notes:

- 1. Pin 1 visual index feature may vary, but must be located within the hatched area.
- 2. § Significant Characteristic
- Dimension D does not include mold flash, protrusions or gate burrs, which shall not exceed 0.15 mm per end. Dimension E1 does not include interlead flash or protrusion, which shall not exceed 0.25 mm per side.
- 4. Dimensioning and tolerancing per ASME Y14.5M
  - BSC: Basic Dimension. Theoretically exact value shown without tolerances. REF: Reference Dimension, usually without tolerance, for information purposes only.
- 5. Datums A & B to be determined at Datum H.

Microchip Technology Drawing No. C04-065C Sheet 2 of 2

#### 14-Lead Plastic Small Outline (SL) - Narrow, 3.90 mm Body [SOIC]

**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



**RECOMMENDED LAND PATTERN** 

Units		MILLIMETERS		
Dimension Limits		MIN	NOM	MAX
Contact Pitch	E		1.27 BSC	
Contact Pad Spacing	С		5.40	
Contact Pad Width	Х			0.60
Contact Pad Length	Υ			1.50
Distance Between Pads	Gx	0.67		
Distance Between Pads	G	3.90		

#### Notes:

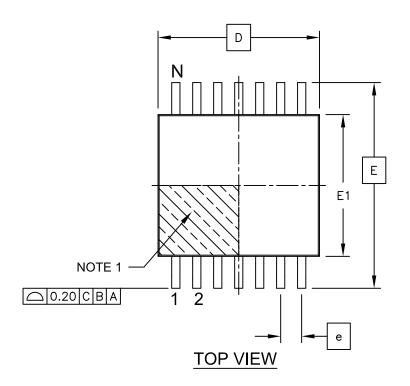
1. Dimensioning and tolerancing per ASME Y14.5M

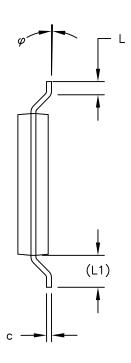
BSC: Basic Dimension. Theoretically exact value shown without tolerances.

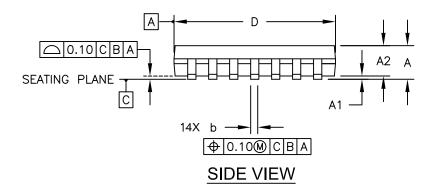
Microchip Technology Drawing No. C04-2065A

#### 14-Lead Plastic Thin Shrink Small Outline (ST) - 4.4 mm Body [TSSOP]

**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



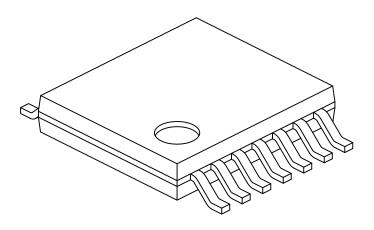




Microchip Technology Drawing C04-087C Sheet 1 of 2

#### 14-Lead Plastic Thin Shrink Small Outline (ST) - 4.4 mm Body [TSSOP]

**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



	Units		MILLIMETERS		
Dimension Limits		MIN	NOM	MAX	
Number of Pins	N	14			
Pitch	е	0.65 BSC			
Overall Height	Α	=	-	1.20	
Molded Package Thickness	A2	0.80	1.00	1.05	
Standoff	A1	0.05	-	0.15	
Overall Width	E	6.40 BSC			
Molded Package Width	E1	4.30	4.40	4.50	
Molded Package Length	D	4.90	5.00	5.10	
Foot Length	L	0.45	0.60	0.75	
Footprint	(L1)	1.00 REF			
Foot Angle	φ	0°	-	8°	
Lead Thickness	С	0.09	-	0.20	
Lead Width	b	0.19	-	0.30	

#### Notes:

- 1. Pin 1 visual index feature may vary, but must be located within the hatched area.
- 2. Dimensions D and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed 0.15mm per side.
- 3. Dimensioning and tolerancing per ASME Y14.5M

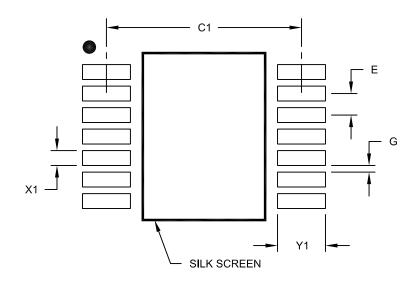
BSC: Basic Dimension. Theoretically exact value shown without tolerances.

REF: Reference Dimension, usually without tolerance, for information purposes only.

Microchip Technology Drawing No. C04-087C Sheet 2 of 2  $\,$ 

#### 14-Lead Plastic Thin Shrink Small Outline (ST) - 4.4 mm Body [TSSOP]

**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



#### RECOMMENDED LAND PATTERN

Units		MILLIMETERS		
Dimension Limits		MIN	NOM	MAX
Contact Pitch	Е		0.65 BSC	
Contact Pad Spacing	C1		5.90	
Contact Pad Width (X14)	X1			0.45
Contact Pad Length (X14)	Y1			1.45
Distance Between Pads	G	0.20		

#### Notes:

1. Dimensioning and tolerancing per ASME Y14.5M

BSC: Basic Dimension. Theoretically exact value shown without tolerances.

Microchip Technology Drawing No. C04-2087A

#### APPENDIX A: REVISION HISTORY

#### **Revision E (November 2014)**

The following is the list of modifications:

- Added SPICE Macro Model in the Related Device features section.
- 2. Updated Temperature Specifications table.
- Corrected pin table in Section 3.0, Pin Descriptions.
- 4. Added new Section 5.4, SPICE Macro Model.

#### **Revision D (February 2013)**

The following is the list of modifications:

 Added the Analog Input (V<sub>IN</sub>) parameter in Section 1.0 "Electrical Characteristics".

#### **Revision C (February 2011)**

The following is the list of modifications:

 Replaced the MCP5468 package name with the correct MCP6567 package name on page 1 and in Table 3-1.

#### **Revision B (August 2009)**

The following is the list of modifications:

- 1. Added MCP6566U throughout the document.
- 2. Updated package outline drawings.

#### Revision A (March 2009)

· Original Release of this Document.

# MCP6566/6R/6U/7/9

NOTES:

#### PRODUCT IDENTIFICATION SYSTEM

To order or obtain information, e.g., on pricing or delivery, refer to the factory or the listed sales office.

PART NO.	– <u>X</u> <u>/XX</u>	Examples:
Device	 Temperature Package Range	a) MCP6566T-E/LT: Tape and Reel, Extended Temperature, 5LD SC70 package.
Device:	MCP6566T: Single Comparator (Tape and Reel) (SC70, SOT-23)	b) MCP6566T-E/OT: Tape and Reel Extended Temperature, 5LD SOT-23 package.
	MCP6566RT: Single Comparator (Tape and Reel) (SOT-23 only)  MCP6566UT: Single Comparator (Tape and Reel) (SOT-23 only)  MCP6567: Dual Comparator	a) MCP6566RT-E/OT: Tape and Reel Extended Temperature, 5LD SOT-23 package.
	MCP6567T: Dual Comparator (Tape and Reel) MCP6569: Quad Comparator MCP6569T: Quad Comparator (Tape and Reel)	a) MCP6566UT-E/OT: Tape and Reel Extended Temperature, 5LD SOT-23 package.
Temperature Range:	E = -40°C to +125°C	a) MCP6567-E/MS: Extended Temperature 8LD MSOP package.
		b) MCP6567-E/SN: Extended Temperature 8LD SOIC package.
Package:	LT = Plastic Small Outline Transistor (SC70), 5-lead OT = Plastic Small Outline Transistor (SOT-23), 5-lead MS = Plastic Micro Small Outline Transistor, 8-lead SN = Plastic Small Outline Transistor, 8-lead ST = Plastic Thin Shrink Small Outline Transistor, 14-lead	a) MCP6569T-E/SL: Tape and Reel Extended Temperature 14LD SOIC package.
	SL = Plastic Small Outline Transistor, 14-lead	b) MCP6569T-E/ST: Tape and Reel Extended Temperature 14LD TSSOP package.

# MCP6566/6R/6U/7/9

NOTES:

#### Note the following details of the code protection feature on Microchip devices:

- Microchip products meet the specification contained in their particular Microchip Data Sheet.
- Microchip believes that its family of products is one of the most secure families of its kind on the market today, when used in the intended manner and under normal conditions.
- There are dishonest and possibly illegal methods used to breach the code protection feature. All of these methods, to our
  knowledge, require using the Microchip products in a manner outside the operating specifications contained in Microchip's Data
  Sheets. Most likely, the person doing so is engaged in theft of intellectual property.
- · Microchip is willing to work with the customer who is concerned about the integrity of their code.
- Neither Microchip nor any other semiconductor manufacturer can guarantee the security of their code. Code protection does not mean that we are guaranteeing the product as "unbreakable."

Code protection is constantly evolving. We at Microchip are committed to continuously improving the code protection features of our products. Attempts to break Microchip's code protection feature may be a violation of the Digital Millennium Copyright Act. If such acts allow unauthorized access to your software or other copyrighted work, you may have a right to sue for relief under that Act.

Information contained in this publication regarding device applications and the like is provided only for your convenience and may be superseded by updates. It is your responsibility to ensure that your application meets with your specifications. MICROCHIP MAKES NO REPRESENTATIONS OR WARRANTIES OF ANY KIND WHETHER EXPRESS OR IMPLIED, WRITTEN OR ORAL, STATUTORY OR OTHERWISE, RELATED TO THE INFORMATION, INCLUDING BUT NOT LIMITED TO ITS CONDITION, QUALITY, PERFORMANCE, MERCHANTABILITY OR FITNESS FOR PURPOSE. Microchip disclaims all liability arising from this information and its use. Use of Microchip devices in life support and/or safety applications is entirely at the buyer's risk, and the buyer agrees to defend, indemnify and hold harmless Microchip from any and all damages, claims, suits, or expenses resulting from such use. No licenses are conveyed, implicitly or otherwise, under any Microchip intellectual property rights.

#### **Trademarks**

The Microchip name and logo, the Microchip logo, dsPIC, FlashFlex, flexPWR, JukeBlox, KEELoQ, KEELoQ logo, Kleer, LANCheck, MediaLB, MOST, MOST logo, MPLAB, OptoLyzer, PIC, PICSTART, PIC<sup>32</sup> logo, RightTouch, SpyNIC, SST, SST Logo, SuperFlash and UNI/O are registered trademarks of Microchip Technology Incorporated in the U.S.A. and other countries.

The Embedded Control Solutions Company and mTouch are registered trademarks of Microchip Technology Incorporated in the U.S.A.

Analog-for-the-Digital Age, BodyCom, chipKIT, chipKIT logo, CodeGuard, dsPICDEM, dsPICDEM.net, ECAN, In-Circuit Serial Programming, ICSP, Inter-Chip Connectivity, KleerNet, KleerNet logo, MiWi, MPASM, MPF, MPLAB Certified logo, MPLIB, MPLINK, MultiTRAK, NetDetach, Omniscient Code Generation, PICDEM, PICDEM.net, PICkit, PICtail, RightTouch logo, REAL ICE, SQI, Serial Quad I/O, Total Endurance, TSHARC, USBCheck, VariSense, ViewSpan, WiperLock, Wireless DNA, and ZENA are trademarks of Microchip Technology Incorporated in the U.S.A. and other countries.

 $\ensuremath{\mathsf{SQTP}}$  is a service mark of Microchip Technology Incorporated in the U.S.A.

Silicon Storage Technology is a registered trademark of Microchip Technology Inc. in other countries.

GestIC is a registered trademarks of Microchip Technology Germany II GmbH & Co. KG, a subsidiary of Microchip Technology Inc., in other countries.

All other trademarks mentioned herein are property of their respective companies.

© 2009-2014, Microchip Technology Incorporated, Printed in the U.S.A., All Rights Reserved.

ISBN: 978-1-63276-677-9

# QUALITY MANAGEMENT SYSTEM CERTIFIED BY DNV = ISO/TS 16949=

Microchip received ISO/TS-16949:2009 certification for its worldwide headquarters, design and wafer fabrication facilities in Chandler and Tempe, Arizona; Gresham, Oregon and design centers in California and India. The Company's quality system processes and procedures are for its PIC® MCUs and dsPIC® DSCs, KEELOQ® code hopping devices, Serial EEPROMs, microperipherals, nonvolatile memory and analog products. In addition, Microchip's quality system for the design and manufacture of development systems is ISO 9001:2000 certified.



### **Worldwide Sales and Service**

#### **AMERICAS Corporate Office**

2355 West Chandler Blvd. Chandler, AZ 85224-6199

Tel: 480-792-7200 Fax: 480-792-7277 Technical Support:

http://www.microchip.com/

support Web Address:

www.microchip.com

Atlanta Duluth, GA

Tel: 678-957-9614 Fax: 678-957-1455

Austin, TX Tel: 512-257-3370

**Boston** 

Westborough, MA Tel: 774-760-0087 Fax: 774-760-0088

Chicago Itasca, IL

Tel: 630-285-0071 Fax: 630-285-0075

Cleveland

Independence, OH Tel: 216-447-0464 Fax: 216-447-0643

**Dallas** Addison, TX

Tel: 972-818-7423 Fax: 972-818-2924

Detroit Novi. MI

Tel: 248-848-4000

Houston, TX Tel: 281-894-5983

Indianapolis Noblesville, IN Tel: 317-773-8323

Fax: 317-773-5453 Los Angeles

Mission Vieio, CA Tel: 949-462-9523 Fax: 949-462-9608

New York, NY Tel: 631-435-6000

San Jose, CA Tel: 408-735-9110

Canada - Toronto Tel: 905-673-0699 Fax: 905-673-6509

#### ASIA/PACIFIC

**Asia Pacific Office** 

Suites 3707-14, 37th Floor Tower 6, The Gateway Harbour City, Kowloon Hong Kong

Tel: 852-2943-5100 Fax: 852-2401-3431

Australia - Sydney Tel: 61-2-9868-6733 Fax: 61-2-9868-6755

China - Beijing Tel: 86-10-8569-7000 Fax: 86-10-8528-2104

China - Chengdu Tel: 86-28-8665-5511 Fax: 86-28-8665-7889

China - Chongging Tel: 86-23-8980-9588 Fax: 86-23-8980-9500

China - Hangzhou Tel: 86-571-8792-8115 Fax: 86-571-8792-8116

China - Hong Kong SAR Tel: 852-2943-5100 Fax: 852-2401-3431

China - Nanjing Tel: 86-25-8473-2460 Fax: 86-25-8473-2470

China - Qingdao Tel: 86-532-8502-7355 Fax: 86-532-8502-7205

China - Shanghai Tel: 86-21-5407-5533 Fax: 86-21-5407-5066

China - Shenyang Tel: 86-24-2334-2829 Fax: 86-24-2334-2393

China - Shenzhen Tel: 86-755-8864-2200 Fax: 86-755-8203-1760

China - Wuhan Tel: 86-27-5980-5300 Fax: 86-27-5980-5118

China - Xian Tel: 86-29-8833-7252 Fax: 86-29-8833-7256

China - Xiamen Tel: 86-592-2388138 Fax: 86-592-2388130

China - Zhuhai Tel: 86-756-3210040 Fax: 86-756-3210049

#### ASIA/PACIFIC

India - Bangalore Tel: 91-80-3090-4444 Fax: 91-80-3090-4123

India - New Delhi Tel: 91-11-4160-8631 Fax: 91-11-4160-8632

India - Pune Tel: 91-20-3019-1500

Japan - Osaka Tel: 81-6-6152-7160 Fax: 81-6-6152-9310

Japan - Tokyo Tel: 81-3-6880- 3770 Fax: 81-3-6880-3771

Korea - Daegu Tel: 82-53-744-4301 Fax: 82-53-744-4302

Korea - Seoul

Tel: 82-2-554-7200 Fax: 82-2-558-5932 or 82-2-558-5934

Malaysia - Kuala Lumpur Tel: 60-3-6201-9857 Fax: 60-3-6201-9859

Malaysia - Penang Tel: 60-4-227-8870 Fax: 60-4-227-4068

Philippines - Manila Tel: 63-2-634-9065 Fax: 63-2-634-9069

Singapore Tel: 65-6334-8870 Fax: 65-6334-8850

Taiwan - Hsin Chu Tel: 886-3-5778-366 Fax: 886-3-5770-955

Taiwan - Kaohsiung Tel: 886-7-213-7830 Taiwan - Taipei

Tel: 886-2-2508-8600 Fax: 886-2-2508-0102 Thailand - Bangkok

Tel: 66-2-694-1351 Fax: 66-2-694-1350

#### **EUROPE**

Austria - Wels Tel: 43-7242-2244-39

Fax: 43-7242-2244-393 Denmark - Copenhagen

Tel: 45-4450-2828 Fax: 45-4485-2829

France - Paris Tel: 33-1-69-53-63-20 Fax: 33-1-69-30-90-79

Germany - Dusseldorf Tel: 49-2129-3766400

Germany - Munich Tel: 49-89-627-144-0 Fax: 49-89-627-144-44

Germany - Pforzheim Tel: 49-7231-424750

Italy - Milan Tel: 39-0331-742611 Fax: 39-0331-466781

Italy - Venice Tel: 39-049-7625286

Netherlands - Drunen Tel: 31-416-690399 Fax: 31-416-690340

Poland - Warsaw Tel: 48-22-3325737

Spain - Madrid Tel: 34-91-708-08-90 Fax: 34-91-708-08-91

Sweden - Stockholm Tel: 46-8-5090-4654

UK - Wokingham Tel: 44-118-921-5800 Fax: 44-118-921-5820

03/25/14

## **Mouser Electronics**

**Authorized Distributor** 

Click to View Pricing, Inventory, Delivery & Lifecycle Information:

Microchip:

MCP6566RT-E/OT