

GENERAL DESCRIPTION

This document describes the specification for the F1950 Digital Step Attenuator. The F1950 is part of a family of *Glitch-Free*TM DSAs optimized for the demanding requirements of communications Infrastructure. These devices are offered in a compact 4x4 QFN package with 50 Ω impedances for ease of integration into the radio system.

COMPETITIVE ADVANTAGE

Digital step attenuators are used in Receivers and Transmitters to provide gain control. The F1950 is a 7bit step attenuator optimized for these demanding applications. The silicon design has very low insertion loss and low distortion (+65 dBm IP3₁). The device has pinpoint accuracy and settles to final attenuation value within 400 ns. Most importantly, the F1950 includes IDT's *Glitch-Free*TM technology which results in less than 0.6 dB of overshoot ringing during MSB transitions. This is in stark contrast to competing DSAs that *glitch as much as 10 dB* during MSB transitions (see p.10).

- ✓ Lowest insertion loss for best SNR
- ✓ Glitch-Free[™] when transitioning won't damage PA or ADC
- ✓ Extremely accurate with low distortion



APPLICATIONS

- Base Station 2G, 3G, 4G, TDD radiocards
- Repeaters and E911 systems
- Digital Pre-Distortion
- Point to Point Infrastructure
- Public Safety Infrastructure
- WIMAX Receivers and Transmitters
- Military Systems, JTRS radios
- RFID handheld and portable readers
- Cable Infrastructure

PART# MATRIX

Part#	Freq range	Resolution / Range	Control	IL	Pinout
F1950	150 - 4000	0.25/ 31.75	Parallel & Serial	-1.3	PE
F1951	100 - 4000	0.50 / 31.5	Serial Only	-1.2	HITT
F1952	100 - 4000	0.50 / 15.5	Serial Only	-0.9	HITT



IDTF1950

150 MHz to 4000 MHz

FEATURES

- *Glitch-Free*TM, < 0.6 dB transient overshoot
- Spurious Free Design
- 3V to 5V supply
- Attenuation Error < 0.3 dB @ 2 GHz
- Low Insertion Loss < 1.3 dB @ 2 GHz
- Excellent Linearity +65 dBm IP3_I
- Fast settling time, < 400 ns
- Class 2 JEDEC ESD (> 2kV HBM)
- Serial & Parallel Interface 31.75 dB Range
- 4 x 4 mm Thin QFN 24 pin package

DEVICE BLOCK DIAGRAM



ORDERING INFORMATION





7-bit 0.25 dB Digital Step Attenuator ABSOLUTE MAXIMUM RATINGS

 $\begin{array}{l} V_{DD} \text{ to GND} \\ D[6:0], \text{ DATA, CLK,LE, V}_{\text{MODE}} \\ \text{RF Input Power (RF1, RF2) calibration and testing} \\ \text{RF Input Power (RF1, RF2) continuous RF operation} \\ \theta_{JA} (Junction - Ambient) \\ \theta_{JC} (Junction - Case) \text{ The Case is defined as the exposed paddle} \\ \text{Operating Temperature Range (Case Temperature)} \\ \text{Maximum Junction Temperature} \\ \text{Storage Temperature Range} \\ \text{Lead Temperature (soldering, 10s)} \end{array}$

150 MHz to 4000 MHz

-0.3 V to +5.5 V -0.3 V to 3.6 V +29 dBm +23 dBm +50 °C/W +3 °C/W Tc = -40 °C to +100 °C 140 °C -65 °C to +150 °C +260 °C



150 MHz to 4000 MHz

F1950 SPECIFICATION (31.75 dB Range)

Specifications apply at V_{DD} = +3.3V, f_{RF} = 2000MHz, and T_{C} = +25°C, EVkit losses are de-embedded (see p. 17) for spec purposes

Parameter	Comment	Sym.	Min	Typical	Max	Units	
Logic Input High	CLK, LE, DATA, D[6:0], V _{MODE}	VIH	2.3		3.6	V	
Logic Input Low	DW CLK, LE, DATA, D[6:0], V _{MODE}				0.7	V	
Logic Current V _{MODE}		Iıн, Iı∟	-5		+5	μA	
Supply Voltage(s)	Main Supply	Vdd	3.0	3.30	5.25	V	
Supply Current	Total	IDD		0.25	0.5 ¹	mA	
Temperature Range	Operating Range (Case)	Тс	-40		+100	°C	
Frequency Range	Operating Range	FRF	150		4000	MHz	
RF1, RF2 Return Loss	dB(s11), dB(s22)	S11, S22		-22		dB	
Minimum Attenuation	D[6:0] = [0000000]	Amin or IL		1.3	1.9	dB	
Maximum Attenuation	D[6:0] = [1111111]	Амах	32.6	33.0		dB	
Minimum Gain Step	Least Significant Bit	LSB		0.25		dB	
Phase Delta	Phase change A _{MIN} vs. A _{MAX}	ΦΔ		34		deg	
Differential Non-Linearity	Max error between adjacent steps	DNL		0.10		dB	
Integral Non-Linearity	Max Error vs. line (A _{MIN} ref) to 13.75 dB ATTN	INL ₁		0.02	0.30	dB	
Integral Non-Linearity	Max Error vs. line (A_{MIN} ref) to 31.75 dB ATTN	INL ₂		0.27	0.45	dB	
	D[6:0] = [0000000] = A _{MIN}	IP3I1	+60 ²	+63			
	D[6:0] = [0111111] = A _{15.75}	IP3I ₂	+59	+61		dBm	
Input IP3	D[6:0] = [1111111] = A _{MAX}	IP3I₃	+57	+61			
	 P_{IN} = +10 dBm per tone 50 MHz Tone Separation 						
0.1 dB Compression Please note ABS MAX Input power on Page 2	 D[6:0] = [0001010] = A_{2.5} Baseline P_{IN} = 20 dBm 	P _{0.1}		27.5		dBm	
Settling Time	 Start LE rising edge > V_{IH} End +/-0.10 dB Pout settling 15.75 – 16.00 transition 	T _{LSB}		400		ns	
Serial Clock Speed	SPI 3 wire bus	Fclk		20	50	MHz	
Parallel to Serial Setup	SPI 3 wire bus	Α	100			ns	
Serial Data Hold Time	SPI 3 wire bus	В	10			ns	
LE delay from final serial clock rising edge	SPI 3 wire bus	С	10			ns	

SPECIFICATION NOTES:

1 – Items in min/max columns in **bold italics** are Guaranteed by Test

2 - All other Items in min/max columns are Guaranteed by Design Characterization



150 MHz to 4000 MHz

SERIAL CONTROL MODE

Serial mode is selected by floating V_{MODE} (pin3) or pulling it to a voltage > V_{IH} . In serial mode data is clocked in LSB first. Note the timing diagram below.

Note – The F1950 includes a CLK inhibit feature designed to minimize sensitivity to CLK bus noise when the device is not being programmed. When Latch enable is high (> V_{IH}), the CLK input is disabled and DATA will not be clocked into the shift register. It is recommended that LE be pulled high (> V_{IH}) when the device is not being programmed.

SERIAL REGISTER TIMING DIAGRAM: (Note the Timing Spec Intervals in Blue)



SERIAL MODE DEFAULT CONDITION:

When the device is powered up it will default to the <u>Maximum Attenuation</u> setting as described below: Note that for the F1950 in all cases (High or 1) = Attenuation Stepped <u>IN</u>. (0 or Low) = Attenuation Stepped <u>OUT</u>.

Default Register \$	Settings
---------------------	----------

					-		•	
	0	1	1	1	1	1	1	1
•	D7 RSV	D6 мsв	D5	D4	D3	D2	D1	D0 LSB

SERIAL MODE TIMING TABLE:

Interval Symbol	Description	Min Spec	Max Spec	Units
Α	Parallel to Serial Setup Time	100		ns
В	Serial Data Hold Time	10		ns
С	LE delay from final serial clock rising edge	10		ns



PARALLEL CONTROL MODE

The user has the option of running in one of two parallel modes: *Direct Parallel Mode* or *Latched Parallel Mode*.

DIRECT PARALLEL MODE:

Direct Parallel Mode is selected when V_{MODE} (pin 3) is < V_{IL} and LE (pin 16) is > V_{IH} . In this mode the device will immediately react to any voltage changes to the parallel control pins [pins 19, 20, 21, 22, 23, 24, 1]. Use direct parallel mode for the fastest settling time.

LATCHED PARALLEL MODE:

Latched Parallel Mode is selected when V_{MODE} (pin 3) is < V_{IL} and LE (pin 16) is toggled from < V_{IL} to > V_{IH}

To utilize Latched Parallel Mode:

- Set LE < V_{IL}
- Adjust pins [19, 20, 21, 22, 23, 24, 1] to the desired attenuation setting. (Note the device will not react to these pins while LE < V_{IL}.)
- Pull LE > V_{IH}. The device will then transition to the attenuation settings reflected by these pins.

Latched Parallel Mode implies a default state for when the device is powered up with $V_{MODE} < V_{IL}$ and LE $< V_{IL}$. In this case the default setting is MAXIMUM Attenuation.

LATCHED PARALLEL MODE TIMING DIAGRAM: (Note the Timing Spec Intervals in **Blue**)



LATCHED PARALLEL MODE TIMING TABLE:

Interval Symbol	Description	Min Spec	Max Spec	Units
Α	Serial to Parallel Mode Setup Time	100		ns
В	Parallel Data Hold Time	10		ns
С	LE minimum pulse width	10		ns
D	Parallel Data Setup Time	10		ns

Insertion Loss vs. Frequency [Amin]

TYPICAL OPERATING PARAMETRIC CURVES (EVKit loss de-embedded unless otherwise noted)



1800

RF Frequency (MHz)

2200

2600

3000

3400

3800

1400



S₁₁ vs. Attenuation State



Attenuation vs. Freq [T_{CASE} = +25C, 0.75 dB steps]



S₂₂ vs. Frequency [T_{CASE} = +25C, 0.75 dB steps]



S₂₂ vs. Attenuation State



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-40 degC - 3.3 V 25 degC - 5.0 V

-25 degC - 3.3 V -100 degC - 3.3 V

1000

0.0

-0.5

Insertion Loss (dB) -1.2 -5.2 -5.2 -5.2

-3.0

-3.5

-4.0

200

600



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150 MHz to 4000 MHz

TOCs CONTINUED (-2-)

Phase vs. Frequency



Supply Current IDD



Input IP3 [f_{RF} = 1900 MHz]



Phase vs. Attenuation Setting







Compression $[f_{RF} = 2000 \text{ MHz}, \text{ ATTN} = 2.5 \text{ dB}]$





IDTF1950 DATASHEET

150 MHz to 4000 MHz

TOCs CONTINUED (-3-)



DNL [900 MHz]



DNL [2800 MHz]



DNL [450 MHz]



DNL [1900 MHz]



Worst Setting DNL





IDTF1950 DATASHEET

150 MHz to 4000 MHz

TOCs CONTINUED (-4-)



INL [900 MHz]



INL [2800 MHz]





40 degC - 3.3 V

25 degC - 3.3 V

100 degC - 3.3 V

8

INL [1900 MHz]

-0.50 -0.75

-1.00

-1.25

-1.50



12

16

Attenuation Setting (dB)

20

24

28

Worst Setting INL





IDTF1950 DATASHEET

150 MHz to 4000 MHz

TOCS CONTINUED (-5-) [f_{RF} = 900 MHz]



The graphs above show the transient overshoot and settling time performance for both the MSB+ and MSB- cases for the F1950. The device settles very quickly (\sim 400) nsec with benign (\sim 0.5) dB overshoot.

Transient [15.75 to 16.00 (MSB+) Standard DSA]



Transient [15.75 to 16.00 (MSB+) 3.3V F1950] Transient [16.00 to 15.75 (MSB-) 5.0V F1950]



The graphs below show the transient overshoot and settling time performance for a popular competing DSA. *Note the overshoot/undershoot excursion of almost 10 dB* and the very long settling time. For the MSB- case, the settling time is off the scale, ~ 3 usec.

Transient [16.00 to 15.75 (MSB-) Standard DSA]





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150 MHz to 4000 MHz

PIN DIAGRAM (F1950)



* Device is RF Bi-Directional



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150 MHz to 4000 MHz

PACKAGE DRAWING (4x4 24 PIN)





PIN DESCRIPTIONS

Pin #	Pin Name	Pin Function
1	D0	Parallel Control – 0.25 dB attenuation step. Pull high for 0.25 dB Attenuation.
2	Vdd	Main Supply. Use 3.3V or 5V. Current is < 1 mA.
3	VMODE	Pull low for parallel mode. Pull high or leave unconnected for serial mode.
4	GND	Connect directly to paddle ground or as close as possible to pin with thru via.
5	RF1	Device RF input or output (bi-directional). Must AC couple to this pin.
6	GND	Connect directly to paddle ground or as close as possible to pin with thru via.
7	GND	Connect directly to paddle ground or as close as possible to pin with thru via.
8	GND	Connect directly to paddle ground or as close as possible to pin with thru via.
9	GND	Connect directly to paddle ground or as close as possible to pin with thru via.
10	GND	Connect directly to paddle ground or as close as possible to pin with thru via.
11	GND	Connect directly to paddle ground or as close as possible to pin with thru via.
12	GND	Connect directly to paddle ground or as close as possible to pin with thru via.
13	GND	Connect directly to paddle ground or as close as possible to pin with thru via.
14	RF2	Device RF input or output (bi-directional). Must AC couple to this pin.
15	GND	Connect directly to paddle ground or as close as possible to pin with thru via.
16	LE	Latch Enable. Serial Data latched into active register on rising edge.
17	CLK	Serial Clock Input
18	DATA	Serial Data Input
19	D6	Parallel Control – 16 dB attenuation step. Pull high for 16 dB Attenuation.
20	D5	Parallel Control – 8 dB attenuation step. Pull high for 8 dB Attenuation.
21	D4	Parallel Control – 4 dB attenuation step. Pull high for 4 dB Attenuation.
22	D3	Parallel Control – 2 dB attenuation step. Pull high for 2 dB Attenuation.
23	D2	Parallel Control – 1 dB attenuation step. Pull high for 1 dB Attenuation.
24	D1	Parallel Control – 0.5 dB attenuation step. Pull high for 0.5 dB Attenuation.
EP	Exposed Paddle	Connect to Ground with multiple vias for good thermal relief.



150 MHz to 4000 MHz

EVKIT SCHEMATIC

The diagram below describes the recommended applications / EVkit circuit:





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150 MHz to 4000 MHz

7-bit 0.25 dB Digital Step Attenuator

EVKIT OPERATION (Email: <u>RFsupport@IDT.com</u> to request an EVkit, Serial Control HW/SW, or TRL cal board)

The picture and graphic below describe how to operate the EVkit





150 MHz to 4000 MHz

EVKIT BOM

F1950 BOM Rev 02 PCB Rev 01	

ltem #	Value	Size	Desc	Mfr. Part #	Mfr.	Part Reference	Qty
1	1000pF	0402	CAP CER 1000PF 50V C0G 0402	GRM1555C1H102JA01D	MURATA	C13,14	2
2	10nF	0402	CAP CER 10000PF 16V 10% X7R 0402	GRM155R71C103KA01D	MURATA	C2,12	2
3	0.1uF	0402	CAP CER 0.1UF 16V 10% X7R 0402	GRM155R71C104KA88D	MURATA	C1,11	2
4	Header 2 Pin	TH 2	CONN HEADER VERT SGL 2POS GOLD	961102-6404-AR	ЗМ	J5,7	2
5	Header 4 Pin	TH 4	CONN HEADER VERT SGL 4POS GOLD	961104-6404-AR	ЗМ	J8	1
6	Header 8 Pin	TH 8	CONN HEADER VERT SGL 8POS GOLD	961108-6404-AR	ЗМ	J6	1
7	SMA_END_LAUNCH	.062	SMA_END_LAUNCH (Small)	142-0711-821	Emerson Johnson	J2,3,4	3
8	0	0402	RES 0.0 OHM 1/10W 0402 SMD	ERJ-2GE0R00X	Panasonic	R1-8,12	9
9	зк	0402	RES 3.00K OHM 1/10W 1% 0402 SMD	ERJ-2RKF3001X	Panasonic	R9-11	3
10	100K	0402	RES 100KOHM 1/10W 0402 SMD	ERJ-2GEJ104X	Panasonic	R13,14	2
11	DIPSwitch	TH 10	8 POSITION DIP SWITCH	KAT1108E	E-Switch	U1	1
12	Digital Step Attenuator		F1950Z	F1950Z	IDT	U2	1
13	PCB		PCB Rev 01	F195XS Evkit Rev 01			1

Total

30

TOPMARKINGS





IDTF1950 DATASHEET

7-bit 0.25 dB Digital Step Attenuator

150 MHz to 4000 MHz

EVKIT THROUGH-REFLECT-LINE (TRL) CALIBRATION

The "Through-Reflect-Line" (TRL) method [1] is used to de-embed the evaluation board losses from the S-parameter measurements of the F1950. This method requires the use of three standards: a through, a reflection, and a line. The TRL method has the advantage over other calibration methods in that it requires only one of these three standards to be well defined.

The TRL through which is used for the F1950 TRL calibration was constructed identically to the evaluation board, minus the DUT and its corresponding length. Therefore, the through corresponds to a precise zero length connection between the input and output reference planes of the DUT. This through satisfies the requirement of the TRL method that one of the three standards be precisely specified.

The TRL reflection standard used is constructed identically to the input and output lines of the evaluation board, with a short placed at the reference plane of the DUT. In accordance with the TRL method's requirements, the actual magnitude and phase were not accurately specified, but the phase was known to within 90 degrees and the TRL reflection standard has a magnitude close to one.

The TRL line standard is identical to the TRL through, but with an additional length of 0.8 inches (2 cm). This satisfies the TRL method's requirement that the TRL be a different length than the TRL through, that it have the same impedance and propagation constant as the through, and that the phase difference between the through and the line be between 20 degrees and 160 degrees. The difference in length yields a phase difference of approximately 20 degrees at 500 MHz, and a phase difference of 160 degrees at 4 GHz.

For characterization of performance from 150 to 500 MHz a separate TRL board with different "Line" length is used.



Standards used for F195x TRL calibration



F1950 evaluation circuit

Engen, G.F.; Hoer, C.A.; "Thru-Reflect-Line: An Improved Technique for Calibrating the Dual Six-Port Automatic Network Analyzer," **IEEE Transactions on Microwave Theory and Techniques**, Volume: 27 Issue:12, pp. 987 – 993, Dec 1979



150 MHz to 4000 MHz

REVISION HISTORY SHEET

Rev	Date	Page	Description of Change
2	2017-Jul-18	2	Corrected Absolute Maximum Supply Voltage.
2	2017-Jul-10	18	Added Revision History Sheet.
1	2013-Jan-15		Corrected Footer
		3	Corrected Maximum Insertion Loss.
		5	Added Parallel Latch Diagram.
0	2012-Nov-04		Initial Release



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